Development of High Power Gyrotron and Power Modulation Technique using the JT-60U ECRF System

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Electron cyclotron range of frequency system of the JT-60U finished operation at the end of August 2008, and improvements toward JT-60SA have been started. In the last two years stable gyrotron oscillation at an output power of 1.5 MW for 1 s was demonstrated, for the first time, using the 110 GHz gyrotron. It was verified that the heat load on the cavity was at an acceptable level with continuous oscillations at 1.5 MW. The absorption power of the collector was also at an acceptable level for the longer pulse oscillation of 5 s. A power modulation technique based on anode voltage modulation was also developed in order to study the effects of modulated Electron Cyclotron Current Drive (ECCD) on Neoclassical Tearing Mode (NTM) stabilization. Modulation frequencies of up to 7 kHz were achieved at output power of 0.8 MW exceeding the previous limit of 3 kHz. Modulated ECCD experiments in synchronization with the NTM were successfully performed with a modulation frequency of around 5 kHz. Development of an accurate synchronization system played an essential role in the experiments that needed a maintained phase between the magnetic probe signal and modulated ECCD in real time. The results provide significant information for further developments that will enhance the overall performance of ECRF systems in the near future.

Keywords: electron cyclotron heating/current drive, high-power gyrotron, power modulation, neoclassical tearing mode, JT-60U

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1. Introduction

Electron Cyclotron (EC) Heating and Current Drive (H/CD) using millimeter-waves would be capable of being a highly efficient and highly localized H/CD. At present plasma confining devices, such as JT-60U, DIII-D and ASDEX-U, EC Range of Frequency (ECRF) systems with mega-watt level of injection power have been installed [1–3]. The International Thermonuclear Experimental Reactor (ITER) requires injection power of at least 20 MW for ECH/ECCD. The JT-60U ECRF system (3 MW for 5 s) commenced operating in 1999, enabling a number of experiments that include profile control, Neoclassical Tearing Mode (NTM) stabilization, plasma start up assistance, heat and momentum transport, and wall cleaning to take place. At the end of August 2008, the system completed operation as JT-60U [4]. Improving the JT-60U ECRF system has already been commenced in order to satisfy the requirements of JT-60SA (1 MW for 100 s per unit).

Over the last ten years R&D has been done on the system in order to enhance the ability of ECH/ECCD with in particular the developments of a high power gyrotron and power modulation technique having been successfully carried out over the past two years. Increasing the output power per gyrotron includes the advantages that higher power could be injected through the same injection port (at the torus) and that the cost and time needed for fabrication, installation and maintenance could be reduced. However, the output power of a gyrotron has been limited to around 1 MW, except for oscillations with a pulse length of 100 ms or less at 1.5 MW or more. The present experiments require pulse duration of at least 1 s. Therefore the development of a high power gyrotron incorporating long pulse capability is an essential issue in order to expand the ability of ECH/ECCD.

Suppression of the NTM is also an important role of ECH/ECCD. Theoretical and simulated results reveal that modulated ECCD in only the magnetic island has the advantage of NTM suppression efficiency compared with that by unmodulated ECCD. However, the effect had not been well confirmed experimentally. In order to study the effect of modulated ECCD in high performance plasmas, such as that in JT-60U, a power modulation technique with a gyrotron of mega-watt power level and a technique of syn-
2. The JT-60U ECRF System

The JT-60U ECRF system consists of four units of power supplies, 1 MW gyrotrons at 110 GHz, evacuated transmission lines with a φ31.75 mm circular corrugated waveguide, and two antennas that can control the beam injection angle into the plasma [1]. Its transmission efficiency, including a matching optics unit, is around 75%. The specified maximum injection power into plasma is 3 MW at a pulse length of 5 s.

Figure 1 gives a schematic view of the power supply, and includes the gyrotron. In this paper k, a, b and c will be used to represent the cathode, anode, body and collector in the gyrotron, respectively. The Main Power Supply (MPS) provides a beam current of up to 65 A with a cathode voltage ($V_k$) of 55 to 60 kV. The electron beam is turned on or off using an Insulated Gate Bipolar Transistor (IGBT) switch within 10 μs. The MPS was originally designed for a lower-hybrid CD system and hence the voltage ripple of the $V_k$ is relatively large for stable gyrotron oscillation. A highly stabilized Body Power Supply (BPS) therefore provides acceleration voltage ($V_{acc}$) between the cathode and the body as the Collector Potential Depression (CPD) operation. The design specifications of the BPS voltage, voltage ripple and current are a maximum of 100 kV, 0.5 kV and 300 mA, respectively. With normal gyrotron operations the $V_{acc}$ is around 80 kV. Here, the CPD voltage ($V_{CPD}$) is given by $V_{CPD} = V_{acc} - V_k$, which is 20 to 30 kV. JT-60U gyrotrons have a triode type Magnetron Injection Gun (MIG), and therefore has an Anode Voltage Divider (AVD) between the body electrode and the anode electrode in order to maintain a suitable pitch factor for the electron beam. The AVD consists of four hundred Zener diodes in series, with up to 80 kV of $V_{ab}$ being provided; each Zener diode provides the $V_{ab}$ of 0.2 kV. The
AVD control unit can be used to modify up to 10 kV of $V_{ab}$ during oscillation with pre-programmed control, and up to 10 kV of $V_{ab}$ can be quickly turned on or off via an optical signal in real time. Each gyrotron has its own BPS and AVD to enable the oscillation parameters to be optimized for each gyrotron independently.

A 1.5 MW dummy load system was newly installed in the gyrotron room to condition the gyrotron and for use in transmission tests. Measuring the rise in temperature of the cooling water of the dummy load system allowed the time averaged output power to be calorimetrically measured while also taking into account transmission efficiency.

### 3. High Power Gyrotron

#### 3.1 Demonstration of 1.5 MW oscillation

Essential issues with high power gyrotron developments include stability of oscillation and the capability of the cavity, collector and other components to withstand the heat load. The time scale with NTM suppression and other plasma behavior in the present experiments ranged from several hundreds of milliseconds to several seconds, thus requiring a gyrotron oscillation with pulse length of at least 1 s. In the frequency range between 28 GHz and 170 GHz gyrotron output power of around 1.5 MW has only been demonstrated in the short pulse region, of 100 ms or less, in showing that high power oscillation was possible. However, space charge neutralization due to ion trapping changes the $V_{ac}$ during the pulse until 100 ms, or sometimes until 1 s, resulting in the oscillation conditions changing. In addition, the cavity radius expands until the heat flux removed by the cooling water reaches the same level as the heat flux due to ohmic loss in the cavity. This expansion also leads to the oscillation conditions changing. Typically, a frequency shift is observed in the initial several hundred milliseconds, which results from the change in oscillation conditions due to the cavity expansion, therefore making the demonstration of stable oscillation with pulse length over 1 s important in verifying the stability of the gyrotron oscillation.

In 2007 a 1.5 MW oscillation for 1 s was demonstrated, for the first time, using the JT-60U ECRF system with an improved 110 GHz gyrotron [5]. Usually the leak current at the anode, $I_a$, is kept low at $I_a < 10$ mA, in obtaining stable oscillation. However, the highest efficiency is obtained at relatively high $I_a$ as described in Ref. [6]. When operating with a high beam current, $I_b$, there is a possibility of the $I_a$ overcurrent failing due to increase in the low quality electron beam component. The above demonstration clarified that $I_a$ be stable between 10 mA and 20 mA, even when $I_b > 60$ A. It was also shown experimentally that the effect of the change in the oscillation condition due to space charge neutralization is sufficiently small to allow a 1.5 MW oscillation. And although the JT-60U ECRF system power supply was used, resulting in slightly lower efficiency than that of the short pulse experiment using the power supply described in Ref. [7], 40 to 45% of efficiency was obtained.

#### 3.2 Cavity and collector temperature measurements

After the demonstration the cavity and collector temperatures were measured in detail in order to investigate the capability of the gyrotron to achieve longer pulse duration at 1.5 MW.

##### 3.2.1 Cavity

Figure 2 gives the cavity temperature with two individual oscillations at 1.5 MW over time, with each of the gyrotron oscillations having been carried out on different days. The sampling time with the thermocouple, which was installed behind the cavity, was 100 ms, resulting in the uncertainty of the measuring time. However apart from that uncertainty the behavior of the temperature increasing with the oscillations was basically the same with good reproducibility being confirmed. A maximum temperature of ~140°C was observed ~0.5 s of the beginning of the oscillation. If the cooling water boils an inter-lock takes place due to water-flow decreasing. However, the pressure of the cooling water of the cavity was estimated to around 0.48 MPa resulting in the boiling temperature of ~150°C, and therefore no boiling was observed. Figure 3 gives the dependence of the temperature rise of the cavity ($\Delta T$) on the power where almost linear dependence of the $\Delta T$ on the output power can be observed. Since the boiling of the cooling water causes the heat transfer coefficient between the cooling water and the cavity to change this also supports that no boiling takes place.

The results thus show that continuous operation with output power of 1.5 MW is possible using the present cavi-
Fig. 3  Power dependence of the cavity saturation temperature. Almost linear dependence is observed. It is thought that the cooling water did not reach boiling point because if the cooling water had boiled the heat removable properties would have changed.

Fig. 4 Collector temperature distribution measured by thermocouples at 1.0 MW for 3 s oscillation. The thermocouples are set in position of the right hand side (R), left hand side (L), output window side (W) and the opposite side (M). The number of thermocouples differs on each side.

3.2.2 Collector

In the previous experiment the collector temperature was only measured at one place (the middle part of the collector, labeled R5) given in Fig. 4 in Ref. [5]. The measured collector temperature did not saturate within 1 s.

In order to investigate long pulse capability from the viewpoint of the collector heat load, detailed measurements of the collector temperature distribution at

\[ \sim 1.5 \text{ MW}, \sim 1.0 \text{ MW} \text{ and } \sim 0.5 \text{ MW} \] were carried out. The sweeping frequency of the collector sweeping coil was 1.7 Hz with a triangular waveform. The low sweeping frequency causes slightly different temperature distributions every time until the temperature saturated. However, the tendency with the temperature distribution was almost the same in every oscillation and had no power dependence. Figure 4 gives an example of an oscillation at 1.0 MW for 3 s, where, R, L, W and M refer to the right hand side, left hand side, output window side and opposite side of the output window, respectively, and the degree of tilt of the magnetic axis in the gyrotron can be seen to be sufficiently small. The highest temperature was observed \( \sim 300 \text{ mm} \) from the bottom of the collector, or more specifically channel 2 or 3 of the thermocouples, and representing the turning point of the swept electron beam. The distribution is consistent with what was expected.

A long pulse experiment was also carried out in order to investigate the saturation time of the collector temperature with the relatively low power of 0.4 MW for the sake of safety. It should be mentioned that an oscillation of 0.5 MW for 30 s had already been successfully achieved with no problems with the collector [4]. Figure 5 gives the measured temperatures at R3, L3, M3 and W2, respectively, over time, where it can be observed that the collector temperature saturated \( \sim 10 \text{ s} \) from the start of the oscillation. The measured saturation temperatures were 160 to 190\(^\circ\)C. In this case, the power injected into the gyrotron, \( V_k \times I_c \), was \( \sim 1.9 \text{ MW} \) at the end of the pulse. A rough estimate was made that \( \sim 1.5 \text{ MW} \) of power had been absorbed on the collector, assuming the oscillation power to be constant and the power of the stray RF relatively low: less than 100 kW and due to low oscillation power. The absorbed amount of power was also confirmed by evaluating the increasing of the cooling water temperature of the collector when it had saturated.

In addition, oscillations of 1 MW for 5 s were achieved simultaneously using four gyrotron units in 2008 [4], in which the gyrotrons had been fabricated with a common collector design. Although all the gyrotrons produced the output power of 1 MW their operation parameters differed.
The power absorbed by each collector was 1.4 to 1.9 MW at 1 s and the power decreased by ~0.1 MW at the end of the oscillation due to the decreasing $I_c$.

Those operating results support the collector of the JT-60U gyrotron to be capable of absorbing at least 1.5 MW with power continuous oscillation, and at least 1.8 MW with up to a 5 s oscillation. In the above 1.5 MW demonstration the power absorbed by the collector was estimated to ~1.7 MW at 1 s. This was also an acceptable level with a pulse length of up to at least 5 s.

This then reveals that the pulse length of a 1.5 MW oscillation to be capable of being increased to 5 s or more from the viewpoint of the collector heat load as well as the cavity heat load.

4. Power Modulation Technique

One of the important roles of ECH/ECCD is NTM suppression [8, 9]. Theoretical and simulated studies revealed that power modulated ECCD increases the efficiency of NTM suppression when compared to unmodulated ECCD. Discussions were being made on whether to use modulated ECCD with a frequency up to 5 kHz in ITER. Previous modulated ECCD experiments had been performed in tokamaks such as JFT-2M [10] and ASDEX-U [11, 12]. However, the effects of modulation and phasing on NTM suppression have not been fully understood experimentally, especially for $m/n = 2/1$ NTM. Here $m$ and $n$ denote the poloidal mode number and toroidal mode number, respectively. A power modulation technique with the gyrotron oscillation and a technique of synchronizing the NTM needed developing for that reason. The JT-60U experiment requires a modulation frequency of around 5 kHz at the mega-watt power level.

4.1 Anode voltage modulation

Several power modulation methods have been developed and can be classified into three cases according to the power supply used for modulation, and (i) a MPS, (ii) BPS, or (iii) AVD (anode power supply) has been used as a modulator in each case. Gyrotrons with a diode type MIG and without CPD do not have a BPS and AVD, and therefore only the MPS can be modulated using such gyrotrons. When using a CPD gyrotron there are two ways of configuring the BPS. One is used on the JT-60U ECRF system given in Fig. 1, where the BPS provides the voltage between the cathode and the body at a voltage of 70 to 90 kV, while the other is given in Ref. [13], and wherein the BPS provides the voltage between the body and the collector at the voltage of 20 to 40 kV. In both cases the BPS can be used as a modulator by which $V_{\text{acc}}$ is modulated with decreasing $V_{\text{CPD}}$. In the former case the MPS is unavailable for modulation because the acceleration voltage is not changed with modulating the MPS. While with the latter configuration, the MPS can be used as a modulator and it has the advantage of reducing collector heat load. The anode voltage modulation by an AVD is another candidate. This technique has the advantages of the modulated devices being easier to use and cheaper than the others but it can only be used with a triode type MIG. It should be mentioned here that the anode voltage modulation technique provides a rectangular waveform rather than a sinusoidal waveform due to the principle of the modulation, which involves the termination of the oscillation due to undesired pitch factor with the optimized $TE_{\text{m},\text{n}}$ mode.

In the present and near future devices requiring several mega-watt ECCD power and gyrotrons incorporating CPD technology will need to be used. However, the power modulation of a CPD gyrotron at a frequency of more than 5 kHz had not been adequately achieved. Reference [13] refers to a modulation frequency of 50 kHz being achieved by modulating the MPS with an output power of 0.8 MW that was decreased to 0.1 MW through modulation. This modulation technique, however, has the disadvantage of requiring expensive regulated power supplies for both the MPS and the BPS, as described in Ref. [13]. In addition, if the MPS is shared by a number of gyrotrons, they all then get modulated simultaneously. A modulation using the BPS was also demonstrated in Ref. [13] at the same output power and modulation depth. However, the modulation frequency was limited to up to 1.5 kHz due to the slow response time of the power supply. An anode voltage modulation technique has been developed with the JT-60U ECRF system. Power modulated ECH at low frequency, less than 1 kHz, was carried out in investigating heat transport [14]. However, the modulation frequency was limited to 3 kHz.

In order to achieve the higher modulation frequency of 5 kHz using the anode voltage modulation technique some improvements were made to one of the ECRF units (#3). Since the photo-couplers used in the AVD limited the response time of the anode voltage, they were replaced by Field Effect Transistors (FETs). Confirmed was made that the FET could be used at a modulation frequency of over 10 kHz. In addition, the resistance and capacitance of the BPS and AVD circuit were modified in obtaining a shorter response time and to reduce the peak charging/discharging current at the electrodes. By adjusting the operating parameters of the gyrotron, a modulation frequency of 7 kHz with an output power of 0.8 MW was successfully achieved, exceeding the previous limit of 3 kHz. Figure 6 gives an example waveform at 5 kHz modulation with the dummy load system. In this case the phase and duty cycle were not controlled in real-time, as described in the following subsection. Typically, modulated voltage between the anode and the body is $\Delta V_{\text{an}} = 3$ to 4 kV. Improving the other ECRF unit (#2) also confirmed that the same modulation frequency and power could be achieved using the same approach. This improvement enabled ECRF power to be injected into the plasma at 0.6 MW for 2 s or 1.2 MW for 1 s utilizing two units with a modulation frequency of up to 7 kHz.
4.2 Synchronizing system

From the technical side there were difficulties with achieving accurately phased modulation synchronized with the NTM as well as obtaining a higher modulation frequency. For instance, a delay time of 20 to 50 µs can be neglected when the modulation frequency is less than 1 kHz, which corresponds to 1 ms per cycle. However, it cannot be neglected in studying the effect of phasing in NTM suppression experiments at a frequency of 5 kHz, which corresponds to 200 µs per cycle, therefore making it worthwhile providing details of those difficulties and an example of the real-time synchronizing system newly developed using the JT-60U ECRF system. Some of the following technical difficulties are common to other modulation techniques too.

Modulation frequency \( f_{\text{mod}} \) can be controlled using measured frequency \( f_{\text{meas}} \), obtained in a real-time measurement. However, even with slight measurement errors of \( \Delta f = f_{\text{meas}} - f \sim 0.1 \) kHz, the errors that accumulate during lots of pulses result in a significant phase error. Here, \( f \) is the exact value of the NTM frequency. The phase error per cycle can be simply described using:

\[
\Delta \Phi = 360^\circ \times \left( \frac{1}{f + \Delta f} - \frac{1}{f} \right) f
\]

\[
= 360^\circ \times \left( \frac{1}{1 + \Delta f/f} - 1 \right). 
\]

(1)

The phase error during \( N \)-th pulses is therefore:

\[
\Delta \Phi^N = N \times \Delta \Phi.
\]

(2)

For instance, \( f = 5 \) kHz, \( f_{\text{meas}} = 5.1 \) kHz results in \( \Delta \Phi^N \sim -7^\circ \times N \), where the minus sign means that the phase of the modulation is faster than that of the NTM signal. And even if real-time control is achieved within 1 ms \( N \) reaches five and thus \( \Delta \Phi^N \sim -35^\circ \), therefore making it necessary that the modulation frequency is not controlled by the measured frequency. By producing each of the modulation signals with respect to each cycle of the NTM the aforementioned accumulation can then be eliminated. However, as \( f \) varies with the rotation velocity of the target plasma the frequency has to be measured in maintaining the duty cycle and phase.

In addition, other difficulties exist with maintaining the duty cycle and phase, even if \( f \) is constant, and the frequency measurement dose provide the correct value. In the modulation test described in Fig. 6 a modulation signal of a rectangular waveform at a constant frequency (5 kHz) and constant duty cycle (on: 0.3; 0.7), \( V_{\text{ref}} \), was provided by the Function Generator (FG) as a reference for the modulation timing. In Fig. 6, the delay time of the diode signal, which corresponds to the oscillated power, is revealed. The delay times were seen to vary during an oscillation. An example of the duty cycle of the diode signal with a modulated oscillation at 5 kHz is provided in Fig. 7 (without RS). In this case, although the duty cycle of \( V_{\text{ref}} \) was adjusted in order to obtain the duty cycle of the diode signal of 0.5 at the start of the modulation, the duty cycle of the diode signal decreased from 0.5 to 0.35 during the oscillation. Those effects cause variations in the phase and duty cycle during an oscillation, thus reducing the effective injection power into the magnetic island, which then causes difficulty with studying the effect of modulated ECCD in NTM stabilization.

In order to achieve modulated ECCD with an accurate phase and duty cycle synchronizing with the NTM rotation, a real-time synchronizing system was developed.
Fig. 8 Schematic view of the newly developed synchronization system. Set values for FG-B are controlled every 50 ms using a PC according to the measured frequency obtained by a digital frequency counter every 20 ms. The phase and duty are preset using the control software on a PC.

Fig. 9 Timing chart of the modulation signals. The numbers on the left hand side correspond to the numbers given in Fig. 8 while RF corresponds to the expected oscillation timing of the gyrotron.

Figure 8 gives a schematic diagram of the system while Fig. 9 gives a timing chart of the control signals. To reference the NTM rotation a magnetic probe signal is used. A TTL level signal is produced if the probe signal is beyond the threshold (0 V). The TTL signal is used in two ways: measuring the NTM frequency for maintaining the phase and duty cycle, and providing the accurate modulation cycle. With the former the signal is enumerated using digital frequency counter at 10 ms of bandwidth, thus providing ±0.1 kHz of frequency resolution every 20 ms. With the latter the signal is used to trigger one of the FG channels (FG-B1). The response time of the threshold circuit causes a little delay time ($t_c$). The FG-B1 then produces the other TTL level signal that includes the delay time of $t_{delay}$. The pulse duration of the signal is sufficiently short (< 10 μs) for the FG-B1 to ready the next trigger signal. In principle the number of generated modulation pulses is exactly the same as that of the NTM enables the accumulated phase error to be eliminated. The TTL pulse produced by the FG-B1 is used to trigger the other FG-B channel (FG-B2).

A single modulation pulse for the AVD control unit is then produced by the FG-B2 at a duty cycle of $t_{duty}$. In order to obtain the target values of the phase ($\Phi_s$) and duty cycle ($R_s: 0 \sim 1$), $t_{delay}$ and $t_{duty}$, used in setting the FGs values, are determined using the following procedure with control software on a PC. The RF power decreases/increases at $T_d/T_u$, which can be described as follows:

$$T_d = t_c + t_{delay} + t_d + (\Delta t_c + \Delta t_d),$$  \hspace{1cm} (3)

$$T_u = t_c + t_{delay} + t_u + (\Delta t_c + \Delta t_u).$$  \hspace{1cm} (4)

Where $t = 0$ denotes when the magnetic probe signal passes beyond the threshold, $t_c$, $t_d$, and $t_u$ are assumed to have been experimentally evaluated beforehand, $\Delta t_c$, $\Delta t_d$, and $\Delta t_u$ are the calibration errors for $t_c$, $t_d$, and $t_u$, respectively. If $\Delta t_c = \Delta t_d = \Delta t_u = 0$ is assumed $t_{duty}$ and $t_{delay}$ can be obtained as follows:

$$t_{duty} = \frac{1}{f_{meas}}(1 - R_s) + t_d - t_u,$$  \hspace{1cm} (5)

$$t_{delay} = \left(\frac{\Phi_s + 180^\circ}{360^\circ} + \frac{1}{2} R_s\right) - \left(\frac{T_u - T_d}{f_{meas}}\right).$$  \hspace{1cm} (6)

However, it does include some experimental errors, and therefore the practical duration of the oscillation interruption obtained in experiments, $T_{off}$, is expressed as:

$$T_{off} = \frac{1}{f_{meas}}(1 - R_s) + (\Delta t_u - \Delta t_d).$$  \hspace{1cm} (7)

The practical oscillating duration, $T_{on}$, is expressed as:

$$T_{on} = \frac{1}{f} - T_{off} = \frac{1}{f} - \frac{1}{f_{meas}}(1 - R_s) - (\Delta t_u - \Delta t_d).$$  \hspace{1cm} (8)

The practical modulation duty cycle, $R_{mod}$, and phase, $\Phi_{mod}$ may slightly differ from the target values of $R_s$ and $\Phi_s$. 

037-7
Φs due to the experimental errors.

\[ R_{\text{mod}} = \frac{T_{\text{on}}}{T_{\text{on}} + T_{\text{off}}} \]

\[ = R_\Phi + \frac{\Delta f}{f_{\text{meas}}} (R_\Phi - 1) - f (\Delta t_{\text{u}} - \Delta t_d). \quad (9) \]

\[ \Phi_{\text{mod}} = \Phi_\Phi - \frac{\Delta f}{f_{\text{meas}}} (\Phi_\Phi + 270') \]

\[ + f \left( \Delta t_c + \frac{1}{2} (\Delta t_d + \Delta t_u) \right) \times 360'. \quad (10) \]

Because of the slow change in rotation velocity of plasmas up to ±2 kHz/s, real-time control every 50 ms by
the PC results in a maximum error of ±0.1 kHz/s. Hence
the system can provide \( f_{\text{meas}} \) with a maximum error of
±0.2 kHz just before the frequency is adjusted taking the
frequency resolution of the measurement into account. But
because the typical time scale with the change in NTM fre-
quency experimentally was much longer than the renewal
interval of the system the frequency can be measured more
accurately, at typically less than 0.1 kHz.

In Fig. 7 also provides an example of duty cycle con-
trol using this system, which is labeled RS. The delay times
of \( t_c, t_u \) and \( t_d \) were calibrated before the experiment. In
this case the sinusoidal signal was produced using the FG
rather than the NTM signal. The duty cycle given in Fig. 7
was obtained by evaluating the time at which the practical
diode signal exceeded the threshold and when the signal
fell below the threshold. The sampling time of the diode
signal was 1 µs, which corresponds to a resolution of the
duty cycle of 0.005 at 5 kHz modulation. Figure 10 pro-
vides histograms of the duty cycle with oscillations both
with RS and without RS. The mean value and standard de-

viation of each histogram are 0.43 and 0.032 for without
RS, and 0.51 and 0.013 for with RS, respectively. The dist-
bution of the duty cycle is not a normal distribution, and
hence the standard deviation does not simply express the
duty control error and is just a measure of dispersion. The
real-time synchronizing system did at least work well with
the sinusoidal signal produced by the FG.

### 4.3 Modulated ECCD experiments with
NTM stabilization

Using the newly developed modulated ECCD system
experiments that aiming to clarify the effect of modulation
and phasing on \( m/n = 2/1 \) NTM suppression were carried
out in JT-60U in 2008. In order to demonstrate the effec-
tiveness of the system a modulated EC-wave was injected
into a plasma where the rotation velocity of the plasma var-
ied due to change in combination of the neutral beams. \( \Phi_\Phi \)
and \( R_\Phi \) were set to be 0° and 0.5, respectively. In this case
\( f \) increased from 4.3 kHz to 6.1 kHz during the modulated
ECCD. Figure 11 gives the NTM frequency obtained us-
ing the fast Fourier transformation of the magnetic probe
signal, the diode signal corresponding to the output power,
the phase difference between the probe signal and the diode

![Fig. 10 Histogram of the duty cycles given in Fig. 7. The mean
value and standard deviation with each histogram were
0.43 and 0.032 for without RS, and 0.51 and 0.013 for
with RS, respectively.](image)

![Fig. 11 Experimental results of modulated ECCD. In this dis-
charge, the NTM frequency varied during the modulated
ECCD. The target values of the phase and duty cycle
were 0° and 0.5, respectively. By the real-time synchro-
nizing system the modulation frequency, duty cycle of
the ECCD and the phase between the NTM and ECCD
were successfully maintained until the mode amplitude
decreased.](image)
signal, and the duty cycle evaluated using the diode signal, all over time. As expected the phase and duty cycle could be correctly controlled at $\Phi_{\text{mod}} \sim 0^\circ$ and $R_{\text{mod}} \sim 0.5$. Since the real-time synchronization system could not produce a modulation signal when the mode was too small, the system terminated the gyrotron oscillation at $t \sim 10.3\,\text{s}$. However, once the NTM decay falls below the threshold corresponding to a marginal island width ($W_{\text{marg}}$) the NTM starts spontaneously decaying. As described in Ref. [15], $W_{\text{sat}} / W_{\text{marg}}$ has a value of around 2 with a typical target plasma with $m/n = 2/1$ NTM in JT-60U, where $W_{\text{sat}}$ is the island width when the NTM is saturated before ECCD. Therefore, and in this case, $m/n = 2/1$ NTM was completely stabilized at $t \sim 10.4\,\text{s}$. 

Experiments were then performed with different phases for each plasma discharge. Figure 12 gives one of the clear effects of phasing. In those plasma discharges the discharge scenarios were fixed and the plasma parameters basically the same. The difference, therefore, in the fluctuation amplitude results from the phasing. Stabilization and destabilization at $\Phi_\delta = 0^\circ$ and $\Phi_\delta = 180^\circ$ were clearly observed.

The details of other experiments and discussions can be found in Ref. [16]. Use of the power modulation technique described in this paper clarified the following. The effect of the phase difference on NTM stabilization was found experimentally consistent with a theoretical model of ECCD efficiency. In addition, comparing the decay time of the mode amplitude during modulated ECCD with that during unmodulated ECCD at the same power, it was found that modulated ECCD is at least twice as effective as unmodulated ECCD.

5. Summary and Further Developments

The developments of a high power gyrotron and a power modulation technique took place using the JT-60U ECRF system in order to enhance the capabilities of ECCD.

Using the improved JT-60U gyrotron 1.5 MW of output power was successfully achieved with a pulse length of 1 s. Confirmed was made that the effects of space charge neutralization and the expansion of the cavity could be neglected in obtaining stable oscillation in this case. Moreover, the leak of current at the anode was stable, even though it was the relatively large amount of $\sim 15\,\text{mA}$ resulting in highly efficient oscillation. The results of measuring the temperature of the cavity and the collector revealed that it would be possible to operate the gyrotron with an output power of 1.5 MW for 5 s or more from the viewpoint of heat load on the cavity and the collector.

In addition, a power modulation technique utilizing anode voltage control was developed. The high frequency of 7 kHz exceeding the requirements of JT-60U and also ITER, was successfully achieved at an output power of 0.8 MW. In order to allow synchronization with the NTM a real-time synchronizing system was newly developed, which worked well. It was shown that modulated ECCD more effectively suppresses $m/n = 2/1$ NTM when compared with unmodulated ECCD, and clarified that phase control is important.

A further development will be minimizing stray RFs in the gyrotrons. The JT-60U gyrotrons used at present have a mode converter that consists of a launcher and phase collection mirrors. However, the mode converter is of low efficiency, therefore making the heating of the internal components by stray RFs a critical issue with long pulse operation. In R&D on 170 GHz gyrotrons for use in ITER at Japan Atomic Energy Agency a newly designed mode converter played a significant role in long pulse operation [6, 17]. Improving the mode converter of the JT-60U gyrotron has already started with the aim of 1 MW with a 100 s oscillation that is needed by JT-60SA. The expectation is that the improvement will also prove effective with a long pulse oscillation of 1.5 MW because of reduced heat load on the collector and less stray RFs. The waveform of the current of the collector sweeping coil should also be optimized more.

At present the pulse length of modulated ECCD per gyrotron is limited to 1 s due to abnormal heating of the launcher in the gyrotron. It is thought that undesired oscillation occurs at the launcher when the oscillation of the primary mode at the cavity is terminated. However, another expectation is that the improved launcher will reduce this kind of undesired oscillation, therefore enabling the pulse duration of modulated ECCD to be expanded to a pulse duration limited by the heat load of the collector.

For synchronizing with the NTM, pre-processing of the reference signal of the NTM which is a magnetic probe signal in our case will be required. If $m/n = 3/2$ NTM occurs the magnetic probe provides fluctuations from the NTM of both $m/n = 2/1$ and 3/2. The signal from $m/n = 3/2$ NTM is dominant when $m/n = 2/1$ NTM gets
decreased by modulated ECCD. Edge Localized Mode (ELM) also provides an irregular signal for use with the magnetic probe. In those cases, accurate control of the phase is difficult. And therefore ideas on filtering or using ECE diagnostics as a reference signal will be required for NTM suppression using modulated ECCD in ITER and other devices.

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